

**Random Generators**  
**Autocorrelation Function and Power Spectral Density**

**Problem 1**

Show that with

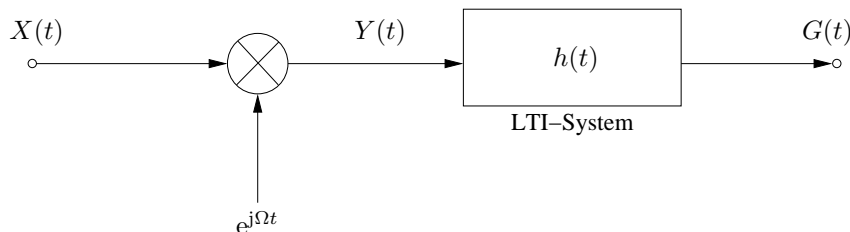
$$Y_1 = \sqrt{-2 \ln X_1} \cos(2\pi X_2) \quad (1)$$

$$Y_2 = \sqrt{-2 \ln X_1} \sin(2\pi X_2) \quad (2)$$

the statistically independent random variables  $X_1$  and  $X_2$  distributed uniformly in  $[0, 1]$  are transformed to “normally” distributed random variables  $Y_1$  and  $Y_2$ .

**Problem 2**

Consider the sample function of a real, stationary and ergodic input signal  $X(t)$  of a transmission system as given in the following.



- 2.1 Calculate the autocorrelation function  $R_{YY}(\tau)$  and power spectral density  $S_{YY}(\omega)$  of  $y(t)$  in terms of  $R_{XX}(\tau)$  and  $S_{XX}(\omega)$ .
- 2.2 Determine the autocorrelation function  $R_{GG}(\tau)$  of the output signal  $g(t)$  in terms of the autocorrelation function  $R_{YY}(\tau)$  of the filter input signal  $y(t)$  and the impulse response  $h(t)$  as a convolution product.
- 2.3 Determine the power spectral density  $S_{GG}(\omega)$  of the output signal  $g(t)$  in terms of the power spectral density  $S_{YY}(\omega)$  of the filter input signal  $y(t)$  and the transfer function  $H(\omega) \leftrightarrow h(t)$ , where  $h(t)$  is the real impulse response.
- 2.4 The system indicated as LTI-System in the plot is a short term integrator with impulse response

$$h(t) = \frac{1}{2T} \text{rect} \left( \frac{t}{2T} - \frac{1}{2} \right) .$$

Write  $R_{GG}(\tau)$  as an integral in terms of the integrand  $R_{XX}(\tau)$  and the independent variables  $\tau$  and  $\Omega$ .

- 2.5 The displayed value of a rms-measuring device at the output of the integrator is proportional to  $R_{GG}(0)$ .

Which integration time  $T$  of the short term integrator is required with the aim to achieve a measured value being proportional to the power spectral density value  $S_{XX}(\Omega)$  if we assume  $R_{XX}(\tau) \approx 0$  for all  $|\tau| > \tau_0$ ?

**Solution of Problem 1**

The probability density function (*pdf*)  $f_{Y_1, Y_2}(y_1, y_2)$  can be generally calculated by

$$f_{Y_1, Y_2}(y_1, y_2) = \frac{f_{X_1, X_2}[x_1(y_1, y_2), x_2(y_1, y_2)]}{\left| \det \frac{\partial(y_1, y_2)}{\partial(x_1, x_2)} \right|}, \quad (3)$$

where the so-called “*Jacobi-Matrix*”  $\frac{\partial(y_1, y_2)}{\partial(x_1, x_2)}$  is defined as

$$\frac{\partial(y_1, y_2)}{\partial(x_1, x_2)} = \begin{pmatrix} \frac{\partial y_1}{\partial x_1} & \frac{\partial y_1}{\partial x_2} \\ \frac{\partial y_2}{\partial x_1} & \frac{\partial y_2}{\partial x_2} \end{pmatrix}. \quad (4)$$

With the four derivatives

$$\frac{\partial y_1}{\partial x_1} = -\frac{2}{x_1} \frac{1}{2\sqrt{-2 \ln x_1}} \cos(2\pi x_2) \quad (5)$$

$$\frac{\partial y_1}{\partial x_2} = \sqrt{-2 \ln x_1} (-\sin(2\pi x_2) \cdot 2\pi) \quad (6)$$

$$\frac{\partial y_2}{\partial x_1} = -\frac{2}{x_1} \frac{1}{2\sqrt{-2 \ln x_1}} \sin(2\pi x_2) \quad (7)$$

$$\frac{\partial y_2}{\partial x_2} = \sqrt{-2 \ln x_1} (\cos(2\pi x_2) \cdot 2\pi) \quad (8)$$

inserted into the determinant yields

$$\begin{aligned} \left| \det \frac{\partial(y_1, y_2)}{\partial(x_1, x_2)} \right| &= \left| \det \begin{pmatrix} \frac{\partial y_1}{\partial x_1} & \frac{\partial y_1}{\partial x_2} \\ \frac{\partial y_2}{\partial x_1} & \frac{\partial y_2}{\partial x_2} \end{pmatrix} \right| \\ &= \left| \frac{\partial y_1}{\partial x_1} \cdot \frac{\partial y_2}{\partial x_2} - \frac{\partial y_1}{\partial x_2} \cdot \frac{\partial y_2}{\partial x_1} \right| \\ &= \left| -\frac{1}{x_1} 2\pi (\cos^2(2\pi x_2) + \sin^2(2\pi x_2)) \right| \\ &= \frac{2\pi}{x_1}. \end{aligned} \quad (9)$$

Since the probability density function  $f_{Y_1, Y_2}(y_1, y_2)$  is depended on  $y_1$  and  $y_2$ , we have to substitute  $x_1$  with the new variables  $y_1$  and  $y_2$ .

With the relation

$$\begin{aligned} y_1^2 + y_2^2 &= -2 \ln x_1 \cos^2(2\pi x_2) - 2 \ln x_1 \sin^2(2\pi x_2) \\ &= -2 \ln x_1, \end{aligned} \quad (10)$$

$x_1$  can be represented in terms of  $y_1$  and  $y_2$

$$\begin{aligned} y_1^2 + y_2^2 &= -2 \ln x_1 \\ \Rightarrow x_1 &= e^{-\frac{1}{2}(y_1^2 + y_2^2)}. \end{aligned} \quad (11)$$

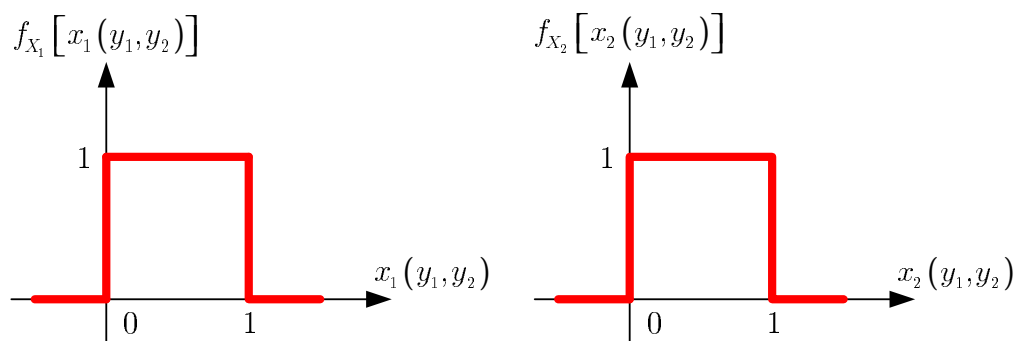
Inserting (11) and (9) into Equation (3) yields

$$f_{Y_1, Y_2}(y_1, y_2) = \frac{1}{2\pi} e^{-\frac{1}{2}(y_1^2 + y_2^2)} f_{X_1, X_2}[x_1(y_1, y_2), x_2(y_1, y_2)]. \quad (12)$$

Since  $X_1$  and  $X_2$  are statistically independent, uniformly distributed in  $[0, 1]$ , we can get

$$\begin{aligned} f_{X_1, X_2}[x_1(y_1, y_2), x_2(y_1, y_2)] &= f_{X_1}[x_1(y_1, y_2)] \cdot f_{X_2}[x_2(y_1, y_2)] \\ &= \begin{cases} 1 & \text{if } 0 \leq x_1(y_1, y_2), x_2(y_1, y_2) \leq 1, \\ 0 & \text{else,} \end{cases} \end{aligned} \quad (13)$$

for arbitrary  $y_1$  and  $y_2$ . The following figure illustrates the probability density functions  $f_{X_1}[x_1(y_1, y_2)]$  and  $f_{X_2}[x_2(y_1, y_2)]$ .



Therefore, we can get

$$\begin{aligned} f_{Y_1, Y_2}(y_1, y_2) &= \frac{1}{2\pi} e^{-\frac{1}{2}(y_1^2 + y_2^2)} \cdot 1 \\ &= \underbrace{\frac{1}{\sqrt{2\pi}} e^{-\frac{1}{2}y_1^2}}_{f_{Y_1}(y_1)} \cdot \underbrace{\frac{1}{\sqrt{2\pi}} e^{-\frac{1}{2}y_2^2}}_{f_{Y_2}(y_2)} \end{aligned} \quad (14)$$

$$= f_{Y_1}(y_1) \cdot f_{Y_2}(y_2), \quad (15)$$

which is the product of the probability density functions of two statistically independent, normally distributed random variables  $Y_1$  and  $Y_2$ .

Solution of Problem 2

2.1

$$\begin{aligned}
R_{YY}(\tau) &= \mathbb{E} \{Y(t)Y^*(t + \tau)\} \\
&= \mathbb{E} \{X(t)e^{j\Omega t} X^*(t + \tau)e^{-j\Omega(t+\tau)}\} \\
&= \mathbb{E} \{X(t)X^*(t + \tau)\} e^{-j\Omega\tau} \\
&= R_{XX}(\tau)e^{-j\Omega\tau} .
\end{aligned} \tag{16}$$

$$\begin{aligned}
S_{YY}(\omega) &= \mathcal{F} \{R_{YY}(\tau)\} = \int_{-\infty}^{+\infty} R_{YY}(\tau)e^{-j\omega\tau} d\tau \\
&= \int_{-\infty}^{+\infty} R_{XX}(\tau)e^{-j\Omega\tau} e^{-j\omega\tau} d\tau \\
&= \int_{-\infty}^{+\infty} R_{XX}(\tau)e^{-j(\omega+\Omega)\tau} d\tau \\
&= S_{XX}(\omega + \Omega) .
\end{aligned} \tag{17}$$

2.2 With  $R_{YG}(\tau) = R_{YY}(\tau) * h(\tau)$ , it can be shown that for a linear time invariant system with the real impulse response  $h(t)$  is valid

$$R_{GG}(\tau) = R_{YY}(\tau) * h(\tau) * h(-\tau) . \tag{18}$$

2.3 We can obtain straightforward

$$S_{GG}(\tau) = S_{YY}(\omega) \cdot H(\omega) \cdot H^*(\omega) = S_{YY}(\omega) |H(\omega)|^2 . \tag{19}$$

2.4 The rect-function is defined as

$$\text{rect}(t) = \begin{cases} 1 & \text{for } |t| \leq \frac{1}{2}, \\ 0 & \text{else.} \end{cases}$$

Figure 1 illustrates the impulse response  $h(t) = \frac{1}{2T} \text{rect}\left(\frac{t}{2T} - \frac{1}{2}\right)$ .

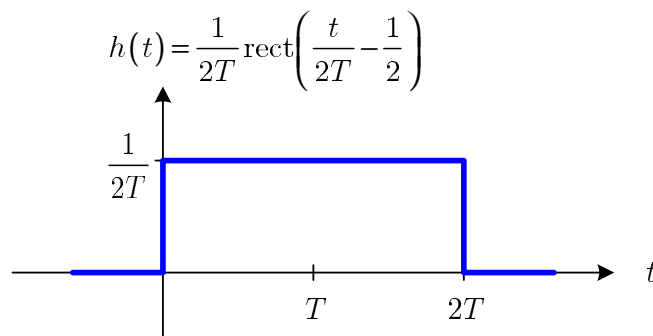


Figure 1: Illustration of the real-valued impulse response  $h(t)$ .

The convolution of the two rect-functions is a triangular function

$$h(\tau) * h(-\tau) = \frac{1}{2T} \left(1 - \frac{|\tau|}{2T}\right) \text{rect} \left(\frac{\tau}{4T}\right) .$$

Figure 2 illustrates the convolution  $h(\tau) * h(-\tau)$ .

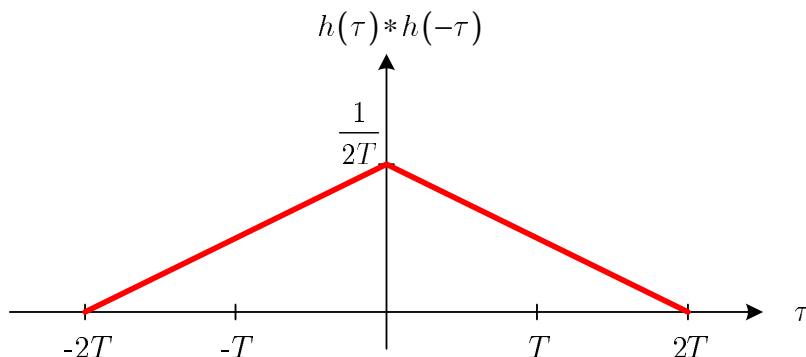


Figure 2: Convolution of  $h(\tau)$  and  $h(-\tau)$ .

Hence, from (16) we have

$$\begin{aligned} R_{GG}(\tau) &= R_{YY}(\tau) * h(\tau) * h(-\tau) \\ &= \frac{1}{2T} \int_{-2T}^{2T} \left(1 - \frac{|u|}{2T}\right) R_{YY}(\tau - u) du \\ &= \frac{1}{2T} \int_{-2T}^{2T} \left(1 - \frac{|u|}{2T}\right) R_{XX}(\tau - u) e^{-j\Omega(\tau - u)} du . \end{aligned} \quad (20)$$

2.5 From 2.4 follows for  $\tau = 0$

$$\begin{aligned} R_{GG}(0) &= \frac{1}{2T} \int_{-2T}^{2T} \left(1 - \frac{|u|}{2T}\right) R_{XX}(-u) e^{-j\Omega(-u)} du \\ &= \frac{1}{2T} \int_{-2T}^{2T} R_{XX}(u) e^{-j\Omega u} du - \frac{1}{2T} \int_{-2T}^{2T} \frac{|u|}{2T} R_{XX}(u) e^{-j\Omega u} du . \end{aligned} \quad (21)$$

The first term of Equation (21) is similar to the Fourier transform of the autocorrelation function  $R_{XX}(\tau)$

$$\begin{aligned} S_{XX}(\Omega) &= \int_{-\infty}^{+\infty} R_{XX}(\tau) e^{-j\Omega\tau} d\tau \\ &= \int_{-\tau_0}^{\tau_0} R_{XX}(\tau) e^{-j\Omega\tau} d\tau , \end{aligned} \quad (22)$$

because we assume that  $R_{XX}(\tau) \approx 0, \forall |\tau| > \tau_0$ .

For  $\frac{\tau_0}{2T} \ll 1$ , (i.e. if the integration time of the short term integrator is much larger than the "correlation time, i.e. the width of the correlation function", the second term in Equation (21) can be neglected because the integration extends only up to  $u = \tau_0$ ) and thus we approximately obtain

$$R_{GG}(0) \approx \frac{1}{2T} S_{XX}(\Omega) . \quad (23)$$

Therefore, for a certain frequency  $\Omega$  we can approximately determine with this circuit apart from the constant factor  $\frac{1}{2T}$  the value of the power spectral density at this frequency.